

Article

# Method for Defining Parameters of Electromechanical System Model as Part of Digital Twin of Rolling Mill

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**Abstract:** Creating digital twins of industrial equipment requires the development of adequate virtual models, and the calculation of their parameters is a complex scientific and practical problem. To configure and digitally commission automated drives, two-mass electromechanical system models are used. A promising area in which to implement such models is the development of digital shadows, namely drive position observers. Connecting virtual models for online data exchange predetermines the tightening of requirements for their parameter calculation accuracy. Therefore, developing accessible techniques for calculating electromechanical system coordinates is an urgent problem. These parameters are most accurately defined by experiments. The contribution of this paper is the proposition of a method for defining the two-mass system model parameters using the oscillograms obtained in the operating and emergency modes. The method is developed for the horizontal stand drives of a plate mill 5000 and is supported by numerical examples. The technique is universal and comprises calculating the rotating mass inertia torques, elastic stiffness and oscillation damping coefficients, and the time constants of the motor air gap torque control loop. The obtained results have been applied to the development of the elastic torque observer of the rolling stand's electromechanical system. A satisfactory coordinate recovery accuracy has been approved for both open and closed angular gaps in mechanical joints. Recommendations are given for the use of the method in developing process parameter control algorithms based on automated drive position observers. This contributes to the development of the theory and practice of building digital control systems and the implementation of the Industry 4.0 concept in industrial companies.



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## 1. Introduction

According to the authors of [1], the current trend in commercial production and monitoring is the digital twin (DT), which is being studied and has shown promising results in facilitating the implementation of the Industry 4.0 concept. The basic idea of using DTs in production is the dynamic digital software interpretation of physical assets and processes [2]. In this context, a digital twin refers to a digital copy of a machine (or its component, e.g., an electromechanical system) or a production line in the course of its development or state change. Therefore, they are defined as follows: “Digital twins are copies of physical production assets which facilitate their monitoring and control” [3]. The application of DTs in production and transport is considered in papers [4–8].

In the metallurgical industry, the objects of DT implementation are the process lines of rolling mills. The created DTs should be used at all life cycle stages (except for disposal), including design, development, virtual commissioning (VC), and asset condition monitoring [9,10]. Therefore, digital twin prototypes (DTPs), instances (DTIs), and aggregates (DTAs) combining DTPs and DTIs into a single complex should be created [11]. Separate areas are developing digital shadows (position observers) of rolling stand electromechanical systems and building control systems on their basis [12,13].

### 1.1. Simulation Models for DTs

As a rule, digital twins are considered in conjunction with simulation models, which are virtual analogues of physical assets [14]. Ref. [5] points to the fact that DTs develop modeling and simulation technologies. In [15], an important conclusion was drawn; it was stated that “computational models used to define the current state of a physical object are the same models that can be used in modeling to predict future states”. This statement is applied in this article for the study of the elastic torque observer on the spindle shaft of the rolling stand.

The problems related to the development and application of digital models for the creation of DTs of industrial systems are considered in [16–19]. Refs. [20,21] note that in recent years, electrical machines powered by a frequency converter have been widely used as torque sources in control systems. Such devices contain complex mechanical transmissions, which may not always be considered rigid. “The operation of such industrial equipment as rolling mills is based on elastic couplings, and a two-mass model of the system is conventionally used for their analysis” [22]. At the same time, “light shafts are used in powerful systems with the significant inertia torques of the electric machine and load. In such conditions, the device model can no longer be considered as a system with a rigid shaft”. Paper [23] considers the dynamics of electromechanical systems with long shafts and elastic couplings. The factors influencing their dynamic behavior, particularly the amplitude of torque oscillations and velocities in a mechanical system, are distinguished. These factors include:

- “parameters of the transmission shaft, i.e., its length and diameter, as well as the inertia torques of the masses attached”;
- “a strategy for controlling an electric motor, while taking into account that the development of this strategy cannot be limited by the dynamics of an electric motor and the synthesis of its control only, excluding other system components”.

In general, [23] confirms the relevance of the selected scientific area in the study of two-mass systems with an elastic shaft. The objectives considered in the paper are prototypes of the developments described in this article.

### 1.2. Multi-Mass Models of Electromechanical Systems in Rolling Mills

A necessary condition for creating and implementing DTs at the aforementioned life cycle stages is the availability of a virtual model describing the object’s properties in operating modes. The electromechanical systems of rolling stands are characterized by the acceleration (with or without metal in the rolls), braking, and load change (shock or smooth) modes determined by the process. In this regard, the most important problem in creating DTs is developing models that reliably describe the object’s behavior in dynamic modes. Scientific papers [24–26] consider developing dynamic models of the electromechanical and mechatronic systems of rolling mills. Ref. [27] emphasizes that when creating DTs, a compromise should be found between the complexity of the mathematical description of variable frequency drives based on Park–Gorev equations and the capabilities of simpler models reliably describing the process, the consideration of which is relevant for solving a specific problem.

When using DTs for the development and VC of drive control systems, in many cases, there is no need to simulate the transients of the “internal” coordinates of motors (current, voltage, magnetic flux). Therefore, it is advisable that simplified simulation

models considering the coordinates that are “external” to the motor are used. These models should reliably describe the elastic properties of the kinematic transmission, the impact of non-linearities (gaps in mechanical joints), the shaft torques, and other non-electrical parameters. To study the electromechanical systems of rolling stands, two-mass (rarely three-mass) models are used [28,29]. In this case, the first and the second masses are, respectively, the motor’s rotor and the roll, which have comparable inertia torques. The masses are generally kinematically linked using a complex-design shaft (spindle) of a bar (slipper) type [30]. It is characterized by elastic properties, manifested during shock load changes, and gaps in the joint mechanisms. Such a two-mass system is adopted for consideration herein. The subject of study is the model parameters to be determined via the experiment.

Ref. [31] emphasized that limiting the dynamic torque is an acute problem for plate and wide-strip hot mill drives. This is because the electromechanical systems of such units operate with a shock load at the moment of workpiece biting (hereinafter, the term “workpiece” refers to a semifinished product between the initial billet–slab and the final product–sheet).

### 1.3. Problem Relevance

The literature overview shows that the “conventional” two-mass model application areas are the study and optimization of process modes [32,33], the synthesis and adaptation of drive controllers [34–36], and the limitation of dynamic loads [37,38]. Due to the recent development of digital technologies, two-mass models are actively used in the following:

- digital twins;
- the development of digital shadows;
- the development of smart control algorithms.

In [39], a two-mass system model is applied to develop drive position observers, which is a promising field. The use of state observers for studying and improving the dynamic performance of two-mass systems is considered in [40–42]. In [43], oscillation and disturbance damping in the system of a two-mass main drive of a rolling mill is studied. The problem is solved using a pole position controller with an improved state feedback and the Kalman filter. Ref. [44] provides a scheme for controlling the speed of rolling mill drives. For the main control loop, a speed controller is proposed; it uses an observer-based state feedback compensator. Ref. [45] proposes a new approach to optimizing observers in a two-mass system and the corresponding choice of a mathematical model; three observers are described for a two-mass electric drive with a flexible shaft.

Theoretically, the analysis and experimental studies of the presented developments show their advantages compared to the conventional PI controller. At the same time, the main disadvantage of the systems with electric drive state observers is the complexity of their practical use. These developments have been tested on special laboratory benches, so no information is given on their industrial adoption. In this regard, the problem of developing relatively simple observers of electric drive coordinates is being solved, which, unlike state observers, do not require a complex mathematical apparatus and control algorithms. Control systems based on them shall be implemented in the software of the logic controllers used in rolling mills.

### 1.4. Methods for Calculating the Parameters of Two-Mass System Models

Papers [46–48] consider the problem of determining the parameters of two-mass models. Ref. [49] highlights that the problem is complicated by the fact that, in many electrical systems, the characteristics of torque transmission on the shaft are nonlinear, and the object parameters are unknown. To solve this problem, an adaptive controller has been developed, and the adequacy of the results obtained has been proven. In [50], a review of the methods used to estimate the parameters of two-mass mechanical systems in electric drives is conducted. An indirect closed-loop method for identifying the parameters of a two-mass system is proposed and compared with other known methods. Based on the validation

results, a conclusion is made on their applicability for estimating the parameters of two-mass mechanical systems. Ref. [51] considers an online indirect parameter identification method that simultaneously estimates six parameters using least squares with an adaptive filter. Its disadvantage is the high computational load that limits its use in real-time identification.

All methods considered are analytical. They conduct the identification of model parameters using computational operations of varying complexity. When making calculations, tabular (nominal) or average data are used, which always differ from the actual parameters of the object. Such a difference can be commensurate with the identification error, which should not exceed 5–10% for electromechanical systems. Thus, the use of the considered methods does not provide the high accuracy required for developing digital twins. Thus, it is necessary to develop a method based on the results of experiments conducted on the object under study. As a result, the actual parameters are “automatically” taken into account, increasing the accuracy of the twin creation (the reproduction of virtual space) of the equipment and processes. The advantage of using this approach is also the ability to conduct experiments at any frequency. This makes it possible to correct the model parameters considering their changes caused by the object’s physical wear and tear.

The authors of [52] rightfully argue that “many researchers implemented the idea of using the observers based on dynamics instead of sensors”. Examples of such observers are considered in the author’s publications [27,53]. Ref. [27] describes a digital observer of the elastic torque on the rolling stand spindle; on its basis, an elastic torque closed control system has been developed. Its technical efficiency has been proven via mathematical models and experiments. However, the study found that the most important factor affecting the reliability of the elastic torque calculation (recovery) in dynamic modes is the electromechanical system model parameter definition accuracy. Obviously, such parameters can be defined most accurately via experiments.

Confirming this conclusion, ref. [54], on the basis of bench tests, substantiates a method for defining the parameters of a multi-mass system with elastic coupling, gaps, and oscillation damping. According to this technique, the motor inertia torque, the motor torque, and the response time of this torque are recognized and corrected. Also, the calculation of the torque response time and the electromechanical time constant of the motor are stipulated for. This paper emphasizes the relevance of the practice-oriented definition of model parameters. Herewith, the literature describes no techniques for calculating the two-mass system parameters using experimental data. This publication aims to fill this gap. The method is developed using the example of the two-mass electromechanical system of the mill stand 5000 described below.

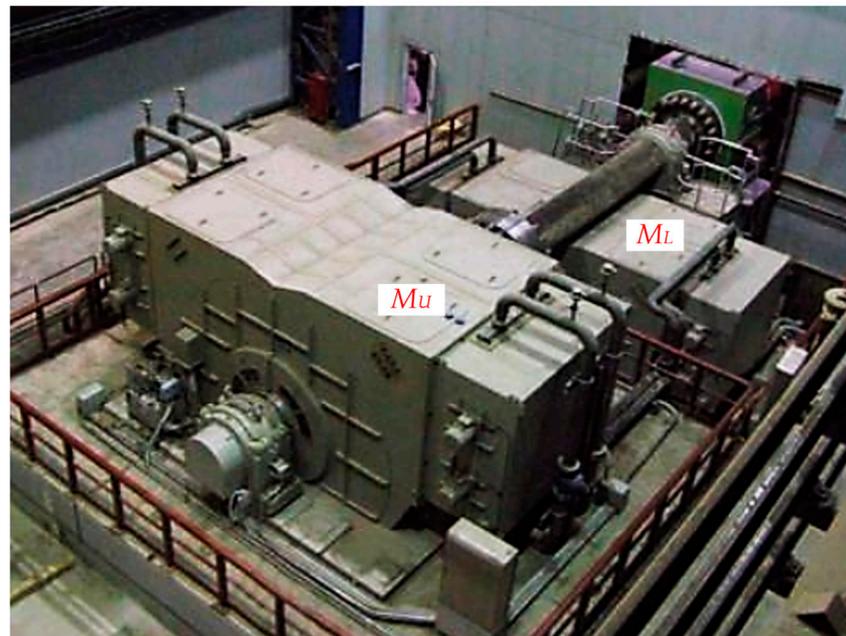
The article content is arranged in the following sequence. Section 2 describes the research object: the electromechanical system of the 5000 mill stand. It also formulates the research objectives. Section 3 presents a method for calculating the parameters of a virtual model of an electromechanical system from the oscillograms of dynamic modes. It gives examples of calculating the model parameters. Section 4 considers the transient processes of the elastic spindle torque, obtained with the help of an observer, and the adequacy of the restored and experimental data is confirmed. Section 5 studies the roll capture under various initial conditions; recommendations regarding the application of the methodology provided in the development of DTs of electromechanical systems are substantiated. Section 6 provides the conclusions drawn regarding the article content and outlines future prospects for research.

## 2. Problem Formulation

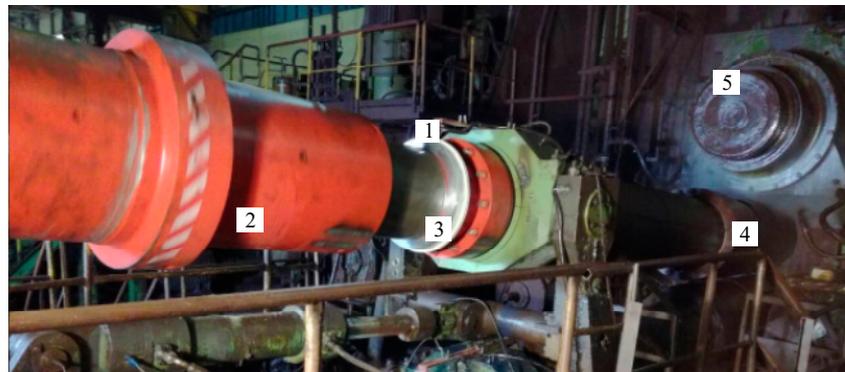
### 2.1. The Mill 5000 Stand Drive Characteristics

The main drives of the upper and lower rolls (UMD and LMD) of the rolling mill under study are arranged individually based on synchronous motors (Figure 1a) with rpm control. Figure 1b shows a photo of spindle 2 connected to upper work roll 4 [27]. Motor rotors are the first “masses” of two-mass electromechanical systems, and the reduced inertia of the work 4 and support 5 rolls is the second “mass”. As shown below, both

masses are comparable in magnitude while there are no devices with a similar inertia in the gears. Therefore, it is fair to assume that the main roll drive lines (main lines) are two-mass electromechanical systems with elastic constraints and angular gaps in the spindle joints. This assumption is valid for most stands of plate and wide-strip hot mills [31,55].



(a)

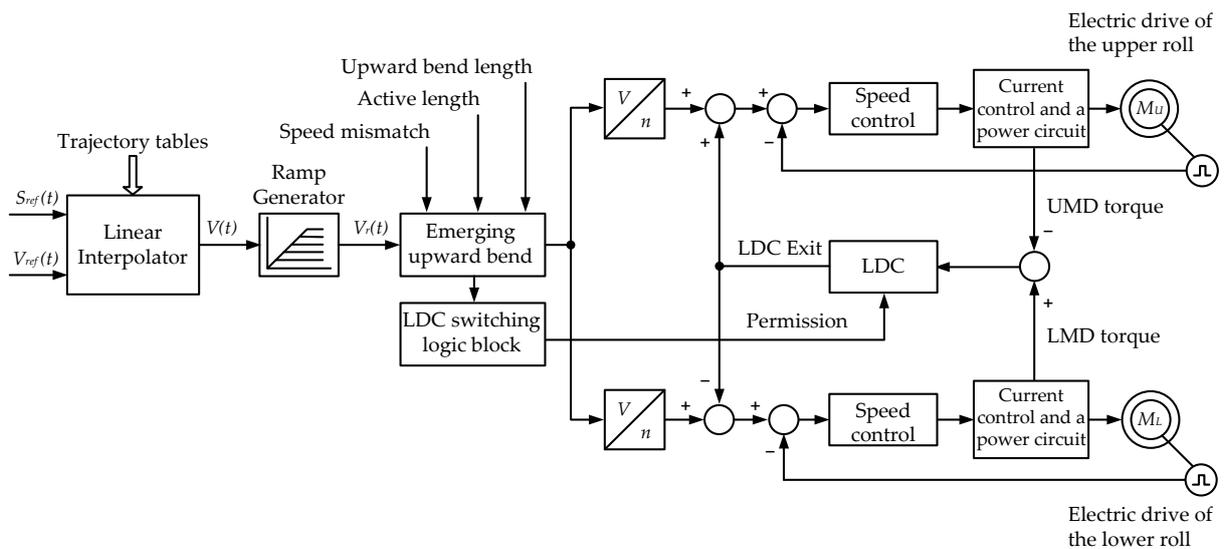


(b)

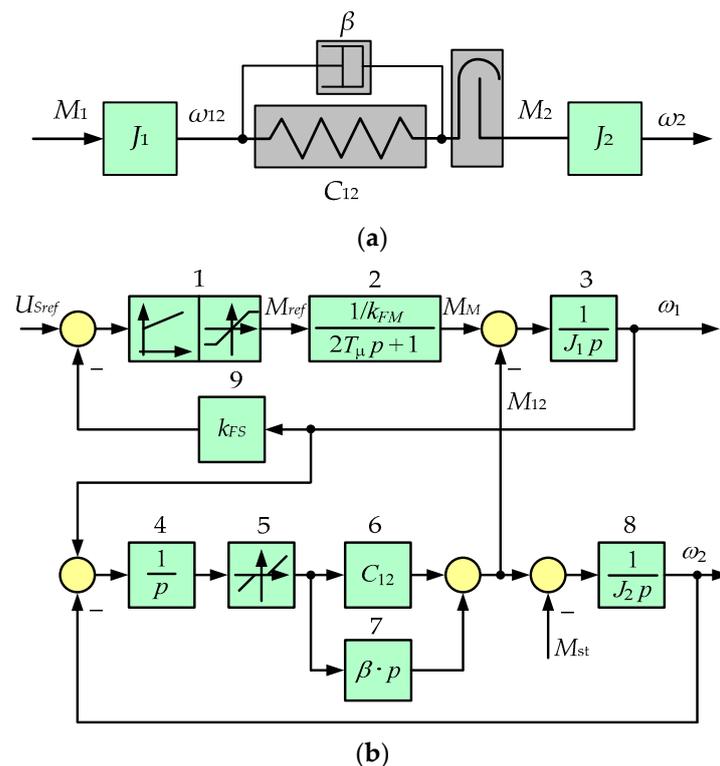
**Figure 1.** The layout of the upper and lower roll motors (a) and the upper roll spindle (b) of the mill 5000 stand: 1 is the sensor; 2 is the spindle shaft; 3 is the telemetry ring; 4 is the spindle head; and 5 is the upper backup roll.

Figure 2 shows the diagram of the system for setting and controlling the UMD and LMD speeds. Synchronous motors  $M_U$  and  $M_L$  are powered by frequency converters; the speed change trajectory is formed by the APCS model according to the criteria of mill performance and by obtaining the specified temperature regime of rolling. The drives are described in more detail in [56].

Figure 3a shows the kinematic diagram of the two-mass system “motor rotor-roll” [27]. Figure 3b shows a simplified diagram of the model of the closed two-loop automated control system (ACS) of the drive speed. In the diagram, blocks 3, 5–7 are typical blocks of a two-mass system model: block 5 simulates gaps in mechanical transmissions, and block 7 defines the natural attenuation of transient oscillations in the mechanical part. Block 2 simulates a multidimensional torque control loop; its transfer function is discussed below in Section 3.



**Figure 2.** Block diagram of the drive control system with implementing the functions of emerging upward bend and load division:  $S_{ref}(t)$  and  $V_{ref}(t)$ —motion trajectory and speed, set as points;  $V(t)$ —drive speed setting generated by the interpolator.



**Figure 3.** Kinematic diagram of the transmission (a) and block diagram of the two-mass electromechanical system (b): 1 is the speed controller; 2 is the speed closed loop; 3 and 8 are the transfer functions of the 1st and 2nd inertial links of the two-mass system (motor and roll); 4 is the integrating link; 5 is the link simulating the angular gap; 6 is the mechanical transmission elasticity coefficient; 7 is the elastic oscillation damping unit; and 9 is the first mass (motor) speed feedback coefficient.

On the diagram,  $T_{\mu}$  is the uncompensated time constant;  $J_1$  and  $J_2$  are the inertia torques of the 1st and 2nd masses;  $C_{12}$  is the mechanical transmission elasticity coefficient;  $\beta$  is the coefficient responsible for natural damping (such as viscous friction);  $M_{ref}$  is the

measured motor torque;  $M_M$  is the electromagnetic torque of the motor;  $M_{12}$  is the spindle elastic torque;  $\omega_1$  and  $\omega_2$  are the motor and roll speeds;  $k_{FS}$  is the first mass speed feedback gain; and  $k_{FM}$  is the motor torque feedback gain in the figure.

The simulated object parameters are provided in Table 1. They are defined by the motor and roll manufacturers and are thus approximate. Under real conditions, the inertia torques are affected by the variable mass of the rolled metal, the inertial properties of the spindle, the thrust bearings, and other mechanical transmission elements. These parameters for the upper and lower rolls differ and cannot be accurately specified when developing the model. Herewith, the reduced mass inertia torques can be defined via experiments with the object, which will provide a high model parameter calculation accuracy. Table 1 also does not contain the magnitude of the angular gap in the spindle joint, which cannot be calculated theoretically but can be defined experimentally. The method and example of measuring the angular gap in the spindle joints of the mill 5000 stand are given in [55] and are not considered herein.

**Table 1.** Source data for defining the two-mass system parameters.

Parameter	Unit of Measure	Value
Motor inertia torque	kg·m <sup>2</sup>	125,000
Work roll weight	kg	63,000
Work roll diameter	kg	1.2
Support roll weight	kg	226,400
Support roll diameter	m	2.3

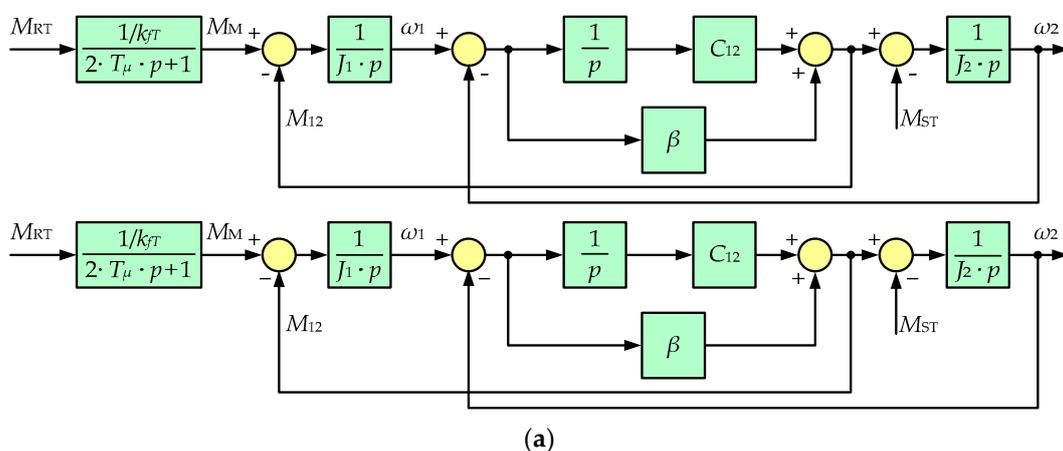
2.2. Two-Mass System Position Observer Characteristic

When developing an observer, non-linear block 5, simulating gaps, and blocks 1 and 9, which implement the speed control loop, are excluded from the diagram shown in Figure 3b. The resulting diagram is shown in Figure 4a and corresponds to the matrix structure shown in Figure 4b [27].

The matrix structure (Figure 4b) corresponds to the system of differential equations in the state space:

$$\begin{aligned}
 \frac{dM_1}{dt} &= -\frac{1}{T_m} M_1 + \frac{1}{T_m K_{oM}} M_{1ref}, \\
 \frac{d\omega_1}{dt} &= \frac{1}{J_1} M_1 - \frac{1}{J_1} M_{12}, \\
 \frac{dM_{12}}{dt} &= C_{12} \omega_1 - C_{12} \omega_2 + \frac{\beta}{J_1} M_1 - \beta \frac{J_1 + J_2}{J_1 * J_2} M_{12} + \frac{\beta}{J_2} M_c, \\
 \frac{d\omega_2}{dt} &= \frac{1}{J_2} M_{12} - \frac{1}{J_2} M_c.
 \end{aligned}
 \tag{1}$$

where  $M_{1ref}$  is the motor torque reference and  $T_m$  is the electromechanical time constant.



**Figure 4.** Cont.

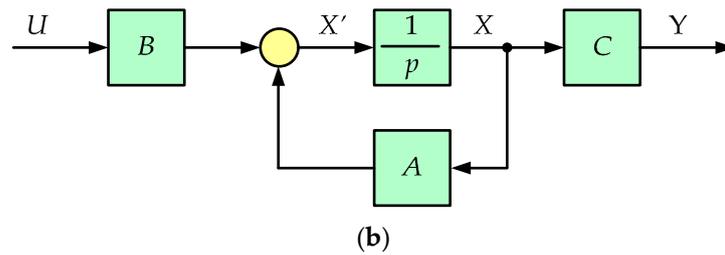


Figure 4. The two-mass system model for the development of an observer (a) and in a matrix form (b).

Based on the above equations, an observer of the elastic torque and the 2nd mass speed has been developed (Figure 5), the detailed information of which is given in [27]. Since the observer is built based upon a model of a two-mass electromechanical system, the accuracy of the online parameter recovery depends on the accuracy of the model coordinate calculation.

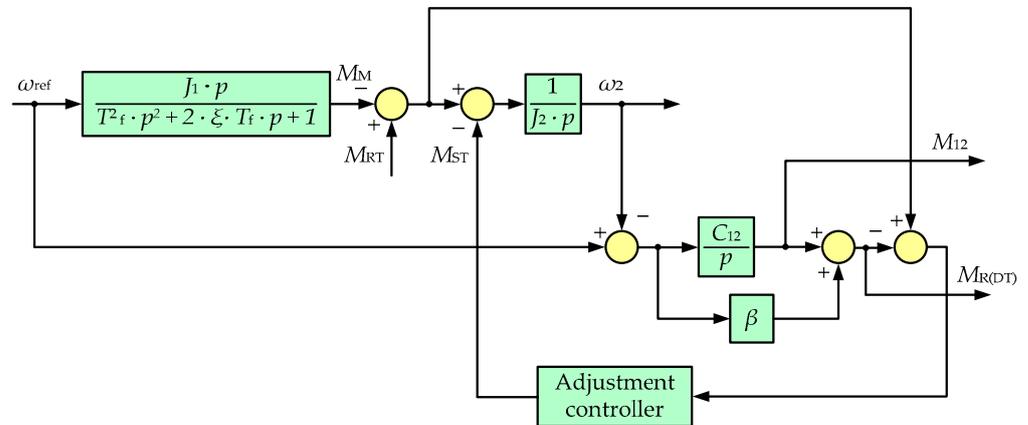


Figure 5. Block diagram of the observer of elastic torque  $M_{R(DT)}$  and the 2nd mass speed.

2.3. Research Objectives

It is difficult to reliably determine the multi-mass system model parameters since some of them cannot be measured directly. These include the reduced inertia torques, which are coefficients characterizing the elastic properties of mechanical transmissions, etc. This requires the development of algorithms for calculating these parameters using the observed coordinates. The developed method should stipulate for the definition of all the model parameters shown in Figure 3b (and, accordingly, Figure 5); its basic aspects are as follows:

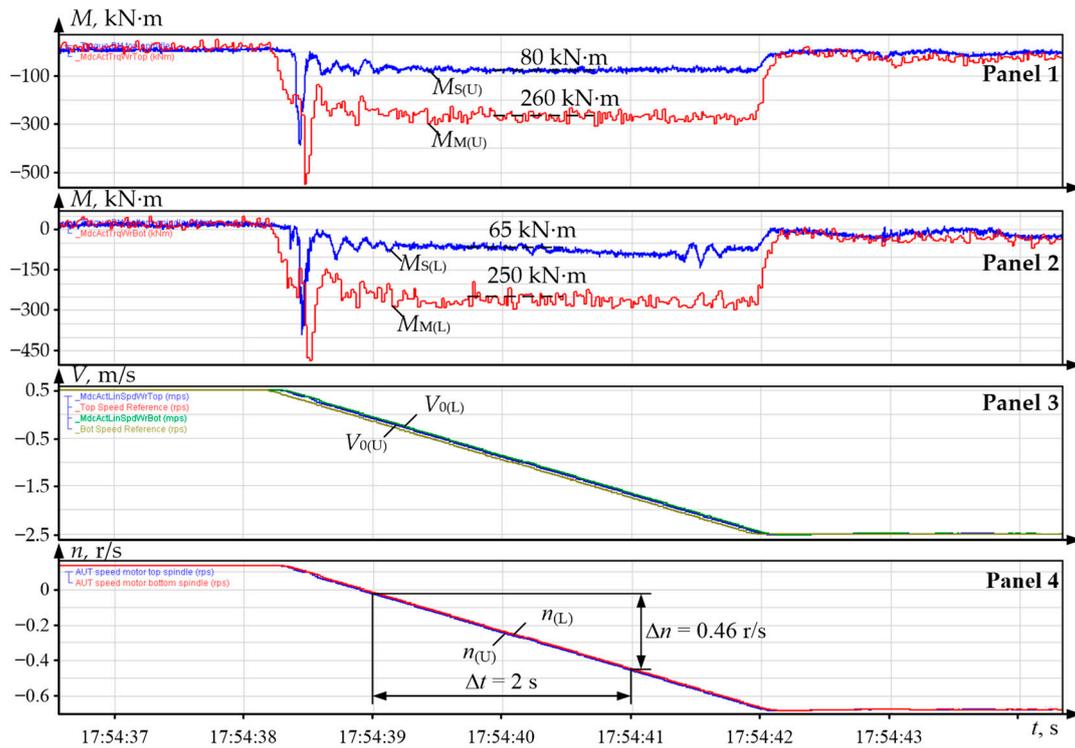
- calculating inertia torques  $J_1, J_2$ ;
- defining the elastic stiffness coefficient  $C_{12}$ ;
- defining the oscillation damping coefficient  $\beta$ ;
- defining the time constants of the transfer function of the torque control loop.

The techniques and examples used to calculate these parameters are discussed below. The calculations are based on the results of processing the oscillograms obtained on the mill 5000 in the operating and emergency modes.

3. Materials and Methods

3.1. Calculating Inertia Torques

To define the inertia torques, the drive constant acceleration experiment was performed. In the example below, they are defined by the acceleration oscillograms in the reverse mode (Figure 6); thus, the speeds and torques are in the negative region. The dependencies of the set and actual speeds in window 3 coincide; therefore, only the set ones are denoted.



**Figure 6.** Acceleration oscillograms of drives without metal in the rolls: window 1— $M_{M(U)}$ ,  $M_{S(U)}$ —upper roll motor and spindle torques; window 2— $M_{M(L)}$ ,  $M_{S(L)}$ —the same for the lower roll; window 3— $V_{0(U)}$ ,  $V_{0(L)}$  speed settings and upper and lower roll speeds (coincide); window 4— $n_{(U)}$ ,  $n_{(L)}$  speeds of UMD and LMD motors.

The motor torques  $M_{M(U)}$  and  $M_{M(L)}$  are measured by the sensors as part of frequency converters. The spindle elastic torques  $M_{S(U)}$  and  $M_{S(L)}$  are measured by the overload monitoring system installed on the mill [57]. Figure 1b above shows the mounting of the elastic torque sensor 1 on telemetric ring 2 fixed on upper roll spindle 3. A similar sensor is installed on the lower roll spindle. The measured torque signals are amplified and transmitted to the data processing and visualization system. They are diagnostic and are not included in the drive control algorithms.

The total two-mass system inertia torque is defined by the dynamic torque  $M_{DYN}$  with the basic drive equation:

$$J_{\Sigma} = \frac{M_{DYN}}{d\omega/dt}. \tag{2}$$

Acceleration is facilitated by the motor air gap torque ( $M_{M(U)}$  or  $M_{M(L)}$ ). In the constant acceleration mode (Figure 6), this torque is constant and depends on the inertia torques of both masses  $J_{\Sigma} = J_1 + J_2$ , e.g., for the upper roll main line,  $J_{\Sigma(U)}$  is defined by the steady-state value of the  $M_{M(U)}$  torque:

$$J_{\Sigma(U)} = \frac{M_{M(U)}}{d\omega/dt} = \frac{260,000}{1.445} = 179,930 \text{ kg} \cdot \text{m}^2.$$

At constant acceleration, the spindle torque ( $M_{S(U)}$  or  $M_{S(L)}$ ) depends only on the second mass (roll) inertia torque  $J_2$ . This is explained by the fact that, in this case, the torque transmitted to the second mass is equal to the spindle’s twisted shaft elastic torque and

does not depend on the first mass inertia  $J_1$ . This allows the second mass inertia torque to be defined for, e.g., the upper roll:

$$J_{2(U)} = \frac{M_{S(U)}}{d\omega/dt} = \frac{80,000}{1.445} = 55,363 \text{ kg} \cdot \text{m}^2.$$

The first mass (rotor) inertia torque is equal to the difference between the  $J_{\Sigma}$  and  $J_2$  values; therefore, for the upper roll engine,  $J_{1(U)} = 124,567 \text{ kg} \cdot \text{m}^2$ .

The dynamic torques defined by the oscillograms (Figure 6) are given in Table 2. It also provides the inertia torque calculation results. The 1st mass inertia torques calculated using the oscillograms differ slightly from the datasheet rate of the motor rotor inertia torque at  $125,000 \text{ kg} \cdot \text{m}^2$ , as specified in Table 1. This confirms the reliability of their experimental definition. The greater magnitude of the 2nd mass inertia torque of the upper roll electromechanical system is explained by the longer upper spindle compared to the lower one. This is confirmed by Figure 1a, where the lower roll motor spindle  $M_L$  is virtually invisible while the motor spindle  $M_U$  is several meters long.

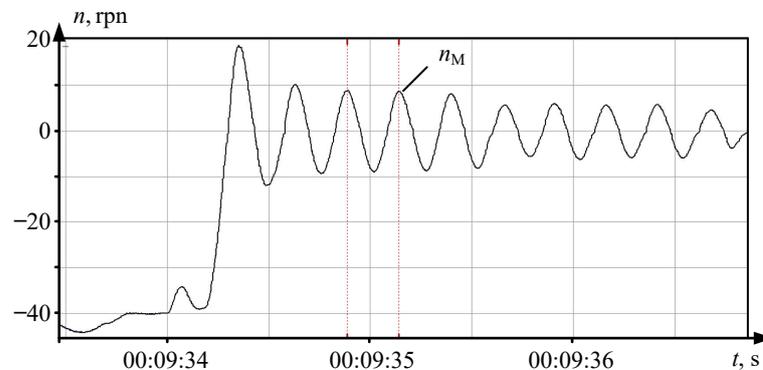
Table 2. The dynamic torques defined by the oscillograms.

Roll Drive	Dynamic Torque, kN·m		Acceleration, rps <sup>2</sup>		Inertia Torque, kg·m <sup>2</sup>		
	Motor	Spindle	rps <sup>2</sup>	rad/s <sup>2</sup>	Total	1st Mass	2nd Mass
Upper	260	80	0.46/2	1.445	179,900	124,537	55,363
Lower	250	65	0.46/2	1.445	173,010	128,027	44,983

The elastic stiffness is defined by the drive emergency shutdown oscillograms. Figure 7a shows the transient speed when the workpiece is stuck in the rolls. In this case, the 2nd mass is fixed, and the 1st mass experiences torsional oscillations due to spindle elasticity. It is known that the torsion pendulum oscillation period is defined by the equation [58]  $T = 2\pi\sqrt{J/C_{12}}$ , from which, with a known 1st mass inertia torque, the spindle stiffness is determined:  $C_{12} = \frac{4\pi^2 J}{T^2}$ . Considering the oscillation period equal to 0.254 s (see Figure 7a), the stiffness is  $C_{12} = 76,489,587 \text{ N} \cdot \text{m}/\text{rad}$ .

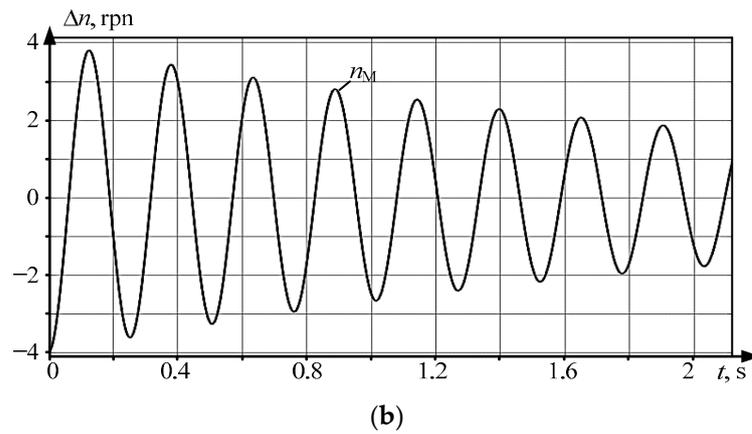
### 3.2. Defining the Oscillation Damping Coefficient

In a real two-mass system, dissipative forces occur, and their impact on damping is considered by the coefficient  $\beta$  in block 7 (Figure 3b). In the pendulum diagram in Figure 8, this block is shown as an amplifier with B factor.

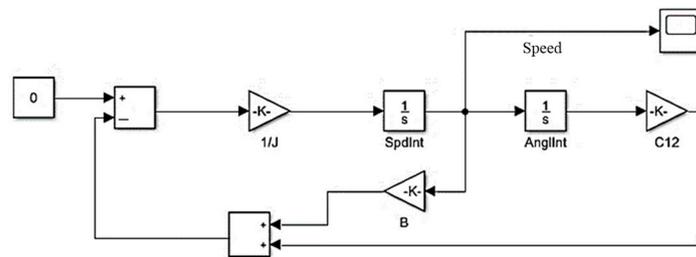


(a)

Figure 7. Cont.



**Figure 7.** Oscillogram (a) and calculated dependence (b) of the motor speed during its emergency shutdown with metal in the rolls (sticking).



**Figure 8.** Two-mass system model in Matlab Simulink to define the damping coefficient.

Using this diagram, the calculated damped speed oscillogram is obtained, as shown in Figure 7b. It shows that the oscillation amplitude decreases by a factor of 2 for approximately 8 speed oscillation periods. The damping coefficient is defined by the logarithmic damping decrement  $\Theta$  [58]:

$$\delta \cdot T = \Theta = \ln \frac{A(t_1)}{A(t_1 + T)}, \tag{3}$$

where  $A(t_1)$  and  $A(t_1 + T)$  are the amplitudes of two successive oscillations, and  $\delta$  is the attenuation coefficient.

The  $T$  value is called the conditional damped oscillation period. Introducing the denotation  $2\delta = \frac{\beta}{J_1}$ , we obtain the following:

$$\beta = \frac{2J_1\Theta}{T}. \tag{4}$$

By substituting the inertia torque  $J_1$  and the parameters  $T$  and  $\Theta$ , defined by the oscillogram in Figure 7a, into Equations (3) and (4), we find that the  $\beta$  value is within 90,000–600,000 N·m·s/rad. Substituting the same parameters into the model, as shown in Figure 8, we find that the coefficient  $\beta \approx 100,000$  N·m·s/rad provides the most accurate coincidence of the curves in Figure 7a,b. In this case, the calculated oscillation period corresponds to the period on the real oscillogram, which confirms the reliability of the model parameter definition.

Indeed, the recommendation to use the emergency mode oscillograms to define the model parameters is not correct. The drive shutdown mode in the presence of metal in the rolls cannot be specially arranged. In the case described, the oscillograms were recorded using a permanent PDA (Process Data Acquisition) system installed on a mill 5000. Currently, such systems operate in all rolling mills, and unfortunately, emergency

modes caused by strip sticking are not uncommon. Therefore, the required oscillograms can be one-time obtained by means of a “passive experiment”.

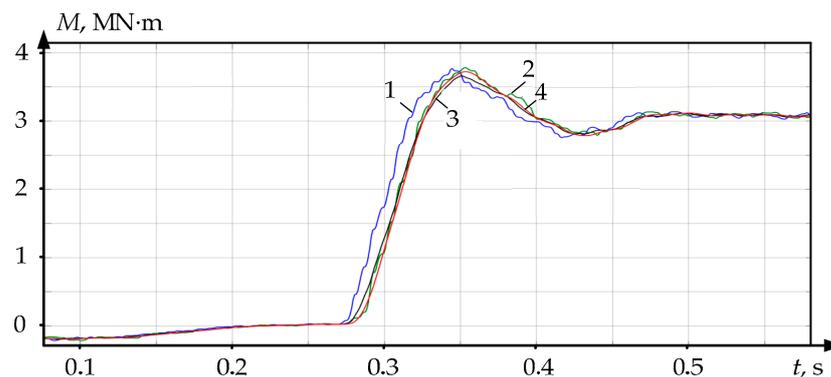
### 3.3. Transfer Function of the Torque Control Loop

Oscillogram in Figure 7a shows that when the drive is turned off, the speed transients have the nature of slowly damped oscillations. Therefore, it is advisable that a closed torque control loop (block 2 in Figure 3b) is expressed as a link with a transfer function:

$$W_{3M}(p) = \frac{1}{T_1^2 \cdot p^2 + 2 \cdot T_1 \cdot \xi \cdot p + 1} \tag{5}$$

This loop can be simulated by first- or second-order links (filters), while defining the time constant  $T_1$  is an important problem. This time constant should be defined by the lag of the actual motor torque curve relative to the torque reference when adjusting the speed setting jump. Such transients are seen in Figure 9, which shows the torque reference curves (speed controller output) and the real (actual) motor torque. When exporting the numerical parameters of these curves to the Matlab package software, the following transfer functions of the approximating links are obtained:

- $\frac{1}{0.008p+1}$ —1st-order filter;
- $\frac{1}{0.007 \cdot 0.007p^2 + 2 \cdot 0.65 \cdot 0.007p + 1}$ —2nd-order filter.



**Figure 9.** Experimental and calculated motor torque transients: 1—setting signal (speed controller output); 2—measured torque; 3, 4—processes at the approximation by the 1st- and 2nd-order filters, respectively.

The figures show that the order of the approximating dependencies has virtually no impact on the set signal processing accuracy: approximating curves 3 and 4 coincide and virtually do not differ from the real torque curve 2. Thus, the torque control loop can be approximated by 1st- or 2nd-order filters with a time constant of 7–8 ms. This simplifies the stand electromechanical system models with virtually no reduction in the simulation result accuracy.

In support, Figure 10 shows the logarithmic amplitude–frequency and phase–frequency characteristics (LAFC and LPFC) of the closed torque control loop under study. When approximating it according to the 1st- and 2nd-order filters, LAFC and LPFC with indices of 1 and 2 were built, respectively. In both cases, the bandwidth is about 100 rad/s, which confirms the possibility of simplification without violating the system properties. It is here assumed that the loop bandwidth is determined by the frequency, after which the signal is attenuated by –3 dB (and further continues attenuating). This magnitude is 100–140 rad/s for both LAFCs. Similar characteristics were obtained in [35].

To test the developed model adequacy for the object under study, the biting and other transients occurring during the rolling cycle were analyzed. The results are given below.

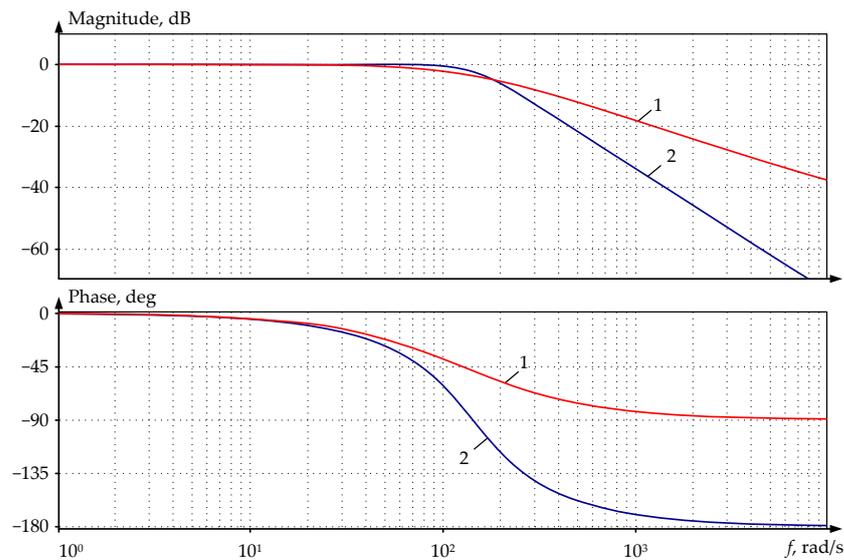


Figure 10. LAFC and LPFC of the motor torque closed loop when approximated by the 1st- and 2nd-order filters.

#### 4. Implementation

The two-mass system digital model parameters, obtained according to the developed method, were used for the VC of the elastic torque observer. To do this, the structure in Figure 5 was implemented in the Matlab Simulink package, and further, the motor speed and torque signals were exported to the model online. This allowed the observer to be configured using not only the parameters calculated in a single point, as in the known drive configuration techniques [59,60], but the dynamic processes occurring in the mill. To estimate the reliability of the results relating to the recovery of the elastic torque by the observer, the recovered and real processes were compared.

Figure 11 shows the transients during the biting, the start of which corresponds to the time instant  $t_1$ . Further, acceleration to a steady rolling speed of 60 rpm is reached at the time instant  $t_4$ . The motor air gap torque  $M_M$  and speed  $n_M$  signals are shown, as obtained from the PDA system. On their basis, the transient process of elastic torque  $M_{R(DT)}$  is recovered using an observer. The real curve of the elastic torque  $M_{R(PDA)}$  is also given, as obtained by the measurement system, the sensor of which is shown in Figure 3b. Dependencies correspond to the case in which, at the moment of biting, the angular gaps in the spindle connections are closed. This is guaranteed by the biting that occurs in the drive acceleration mode in the time interval from the origin of coordinates to the time instant  $t_1$ . Therefore, the dynamic shock caused by closing gaps does not introduce errors in the torque recovery.

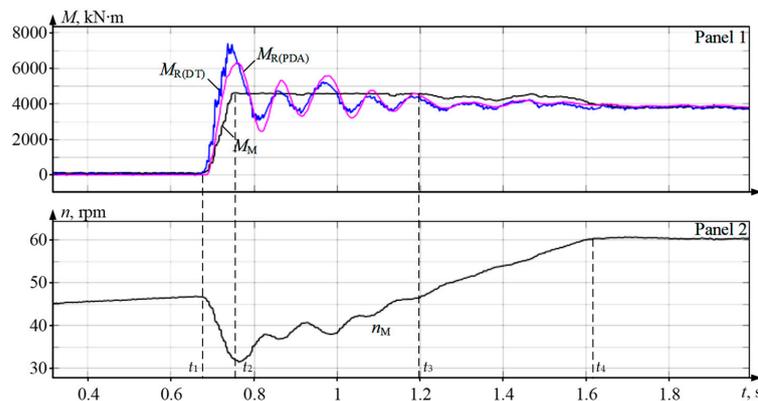
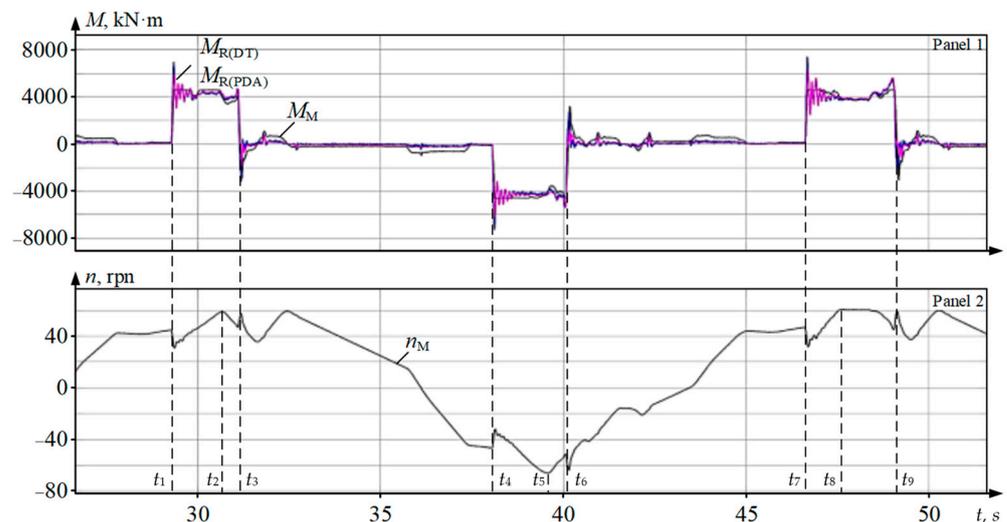


Figure 11. Transients of the upper roll spindle elastic torque at the biting and acceleration.

In the interval  $t_2-t_3$ , the motor torque  $M_M$  reaches the limit of 4200 kN·m, which leads to a loss of motor controllability. In this regard, the elastic torque transient process has the nature of damped oscillations. Transients of the restored  $M_{R(DT)}$  and measured  $M_{R(PDA)}$  elastic torques correspond to each other with an error not exceeding 10%, so the model and the object can be considered adequate. This indirectly confirms the reliability of calculating the two-mass system (and, accordingly, the observer) coordinates using the method described above. This error is explained by the model's inability to accurately consider some factors affecting the torque amplitude, particularly the load increase rate during biting, which depends on the shape of the workpiece's front end and its speed on the roller table.

To summarize the results, Figure 12 shows transients of the same coordinates as Figure 11, obtained for three reverse rolling passes in the time intervals  $t_1-t_3$ ,  $t_4-t_6$ , and  $t_7-t_9$ . The biting occurs at the time instants  $t_1$ ,  $t_4$ , and  $t_7$ , acceleration with metal in the rolls occurs in the intervals  $t_1-t_2$ ,  $t_4-t_5$ , and  $t_7-t_8$ , and the workpiece exit from the rolls occurs at the time instants  $t_3$ ,  $t_6$ , and  $t_9$ . Thus, all transients occurring during the rolling cycle are presented, so these oscillograms allow the accuracy of the elastic torque recovery in all dynamic modes to be estimated.



**Figure 12.** The spindle elastic torque and motor speed transients for three passes.

Based on these oscillograms, the following conclusions were drawn:

1. In all passes, biting occurs with closed angular gaps in the spindle joints.
2. The full coincidence of the recovered  $M_{R(DT)}$  and measured  $M_{R(PDA)}$  torques confirms the reliability of the processes obtained during the VC of the developed observer.
3. Time dependencies in Figures 11 and 12 illustrate the adopted approach to configuring digital twins, according to which algorithms are debugged in Matlab Simulink with subsequent export to PLC software [12]. This facilitates configuration and reduces the VC costs of the electromechanical system.

Similar conclusions were drawn based on the oscillograms obtained for the lower roll drive (not shown here).

The method developed for calculating the virtual model parameters based on oscillograms of dynamic modes is recommended for use in the development of DTs of electromechanical systems. The described model is used to study transients and develop algorithms for limiting dynamic loads in the main lines of rolling stands.

## 5. Results and Discussion

To summarize the results and estimate the impact of angular gaps on the elastic torque recovery accuracy, rolling modes studies were compared for more than 150 workpieces

with a gauge of 9 to 30 mm under various biting conditions. Figure 13 shows the average dependencies of the spindle torque amplitude with open and closed gaps (“gapped” and “gapless” diagrams, respectively). As noted above, the preliminary gap closing was facilitated by the drive acceleration; accordingly, gapped biting was provided by the short-term drive braking. The dependence “observer” of the elastic torque recovered using an observer is also shown here. The numerical values used to build these dependencies are given in Table 3.

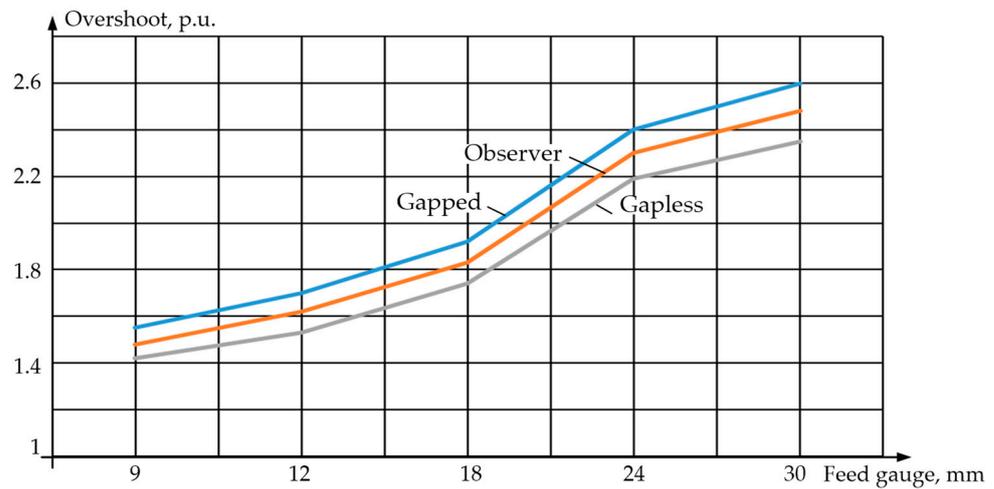


Figure 13. Dependences of the biting torque amplitude, obtained experimentally and recovered by the observer.

Table 3. Calculated biting torque amplitudes.

Gauge	Measured Values				Observer Values				
	$M_{max}$ (Gapped)	$M_{max}$ (Gapless)	$\Delta M_{max}$ (Gap/Gapl.)	%	$M_{max}$ (Obs.)	$ \Delta M_{max} $ (Obs./Gap)	%	$ \Delta M_{max} $ (Obs./Gapl.)	%
mm	p.u.	p.u.	p.u.	%	p.u.	p.u.	%	p.u.	%
9	1.55	1.42	0.13	8.4	1.48	0.07	4.5	0.06	4.2
12	1.7	1.53	0.17	10.0	1.62	0.08	4.7	0.09	5.9
18	1.92	1.74	0.18	9.3	1.83	0.09	4.7	0.09	5.2
24	2.4	2.19	0.21	8.8	2.3	0.1	4.2	0.11	5.0
30	2.6	2.35	0.25	9.6	2.48	0.12	4.6	0.13	5.5

The analysis of the data in the table allows the following conclusions to be made:

1. The recovery of the elastic torque by an observer, whose parameters have been determined using the developed method, is performed with an accuracy acceptable in the study of electromechanical systems. The maximum difference between the measured and recovered gapped values  $|\Delta M_{max}|_{(Obs./Gap)} = 4.7\%$ , while the “gapped” curve parameters exceed those of the “observer” curve at all points. The respective gapless value  $|\Delta M_{max}|_{(Obs./Gapl.)} = 5.9\%$ , but the ‘gapless’ dependence coordinates are less than those on the “observer” curve. Thus, the recovered dependence occupies an intermediate position between the experimental curves. This allows the elastic torque (“observer”) recovery results to be considered acceptable for both the “gapless” and “gapped” cases.
2. The experiments confirm that closing the gaps at the moment of biting significantly increases the elastic torque amplitude (within 8.4–10%). With workpiece gauges of 9 and 30 mm, the amplitude differences are 8.4% (1.55 and 1.42 p. u.) and 9.6% (2.6 and 2.35 p. u.), respectively. This confirms the need for implementing con-

trol systems based on elastic torque observers to reduce the dynamic loads of the electromechanical systems of rolling mills [27].

General conclusion: the developed two-mass system coordinate position observer provided a satisfactory accuracy for parameter recovery and can be used in developing closed-loop elastic torque ACS.

The method proposed for determining the parameters of an electromechanical system model includes conducting the experiments with the engine running or stopped. It can be used for a wide class of electromechanical systems with elastic shafts under constant or variable load. The method is recommended for studying the equipment of any rolling mills, the electromechanical systems of which can be represented as two-mass models. In addition to the plate and wide-strip mills discussed, these include section rolling mills, sheet cold rolling mills, and some types of pipe rolling mills.

As noted above, the method is experimental. Within this approach, the system under study is essentially presented as a “black box”. Thus, there are no restrictions in terms of the complexity of the simulated electromechanical system. The method is applicable to two-mass and three-mass systems with an elastic shaft, which has been experimentally confirmed.

Since the method is not computational and does not involve the use of complex mathematical apparatus, no uncertainties arise in the process of determining the system parameters. The required parameters are defined during experiments, which can be conducted at specified intervals. This makes it possible to take into account changes in the dynamic behavior of the system, particularly regarding the wear of spindle connections. However, practical experience in operating electromechanical systems shows that such changes occur slowly. Thus, the standard service life of rolling mill spindles is 8 years (this is not always achieved). However, there is no need to consider the dynamic changes in the objects under study. These properties highlight the advantages of the experimental approach proposed.

## 6. Conclusions and Future Work

Studying the dynamic mode of the rolling stand electromechanical system at biting is an important problem. When using DTs in the development and virtual commissioning of automated drives, it is advisable to represent it as a two-mass model with an elastic shaft and angular gaps in the joints. This model reduces the quantity of software resources required to create DTs and thus allows DTs to be introduced in the PLC software.

A method for calculating the two-mass system virtual model parameters by using experimental data has been developed. Its key points are as follows:

- calculating the 1st and 2nd mass inertia torques;
- defining the stiffness coefficient of the mechanical transmission shaft;
- defining the oscillation damping coefficient;
- defining the time constants of the torque control loop transfer function.

The method is based on the analysis of oscillograms obtained in dynamic modes. The inertia torque is calculated based on the load-free drive acceleration transients. The elastic stiffness coefficient is defined by the motor emergency shutdown oscillograms with the metal in the rolls, obtained via a passive experiment. Similar oscillograms are used to define the oscillation damping coefficient. The approximation of the torque control loop in the two-mass system model using a filter with a time constant of 7–8 ms is justified. This allows the closed loop for controlling the motor air gap torque to be represented as a first-order link.

The reliability of the elastic torque recovery with satisfactory accuracy was confirmed by comparing the observer VC results and the experimentally recorded oscillograms. The calculation error of this parameter in the biting mode with closed angular gaps does not exceed 5.9%, and when gapped, does not exceed 4.7%. In this case, the largest difference in the elastic torque amplitudes due to the gaps closing at the moment of biting is 10%.

Thus, model validation was defined according to a “conventional” method of comparing the final calculation results with the experimental results. The criteria for determining

the adequacy of the reconstructed and experimental data correspond to the generally accepted error, which is acceptable when modeling industrial electromechanical systems (in most cases, the discrepancy between the calculated and experimental data should not exceed 5%). As follows from the calculations above, the model under consideration provides such accuracy. For verification, this study used a method of comparing the sequence of events in a real system with the events occurring in the model, and that involved the interactive monitoring of the progress of modeling the dynamic modes for the systems under study. Since the two-mass model is generally accepted, no special attention was paid to this issue.

The electromechanical system VC technique has been applied, according to which control systems are configured using models in the Matlab Simulink package. Next, the models and algorithms are exported to the object controller software. The proposed method for configuring the observer is an example of implementing the DT technology at the stages of drive control system development and virtual commissioning. This field is considered in [12,27], and is considered promising regarding the development and improvement of the electromechanical and mechatronic systems of rolling mills.

State observers should form the backbone of rolling mill DTAs. Creating relatively simple object-oriented digital twins will contribute to the development of smart production and the transition of the metallurgical industry towards innovations [61–64]. Virtual models required to improve automated drives and mechatronic systems, including the described adequate two-mass model, facilitate this [65–67].

Building accessible virtual models reduces the materials and time spent on the implementation of digital technologies in operating industrial units. Exporting relatively simple object-oriented DTs to the PLC software facilitates their real implementation, which is the most important IoT problem.

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