



Article Position Output Adaptive Backstepping Control of Electro-Hydraulic Servo Closed-Pump Control System

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Abstract: A nonlinear adaptive backstepping control method was proposed to address the system parameter uncertainty problem in the position control process of an electro-hydraulic servo closed-pump control system. This control strategy fully considers the parameter uncertainty of the nonlinear system and establishes the adaptive rate of the uncertain parameter to adjust the parameter disturbance online in real time, thereby improving the accuracy and robustness of the control system. A pump control system experiment platform was used to verify the feasibility of the controller. The experimental results showed that the proposed control strategy provided a good control effect. The pump control system can be controlled with high precision, with a steady-state control accuracy of ± 0.02 mm, which serves as a good foundation for the engineering application and promotion of the pump control system.

Keywords: pump-control system; position control; backstepping; adaptive backstepping control

1. Introduction

The electro-hydraulic servo control system is vital to hydraulic technology and is one of the basic technical components of modern control engineering [1]. Because of its high load-bearing capacity, dynamics, and precision, it is widely used in manufacturing, national defense, military industry, airspace, and other fields [2–4]. On the basis of its operating principle, electro-hydraulic servo technology can be classified into two categories: pump control and valve control. Traditional electro-hydraulic servo technology is primarily based on valve control systems. However, owing to the development of electro-hydraulic servo technology in the manufacturing field, electro-hydraulic servo technology with high response, high precision, high power-weight ratio, high power, and low energy consumption has become the goal pursued in the development of hydraulic integrated equipment [5]. In recent years, variable frequency drive and servo drive technology have been gradually perfected; therefore, electro-hydraulic servo closed-pump control (hereinafter referred to as pump control) technology has been developed considerably. Compared with the traditional electro-hydraulic servo valve control technology, pump control technology can not only effectively solve the inherent defects of the former, but also offers the advantages of energy efficiency, high power-weight ratio, and environmental friendliness [6]. Therefore, it is preferred by engineers and has become a topic of focus in the field of electro-hydraulic servo research and development.

An electro-hydraulic position servo system is a typical uncertain nonlinear system [7] that exhibits many nonlinear characteristics and model uncertainties [8]. At present, to address the problems due to of transient parameter time variation and nonlinear factors [9-12]



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Copyright: © 2021 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license (https:// creativecommons.org/licenses/by/ 4.0/). in electro-hydraulic position servo system [13–15], researchers worldwide primarily adopt feedback linearization [16], sliding mode control [17–23], adaptive control [24–27], fuzzy control [28,29], and other technologies. As a systematic and structured design method is adopted in backstepping, uncertainties and unknown parameters can be addressed easily in the system [30]. Backstepping can ensure that the final controller exhibit Lyapunov asymptotic stability; therefore, it has been investigated and applied by many scholars.

Kaddissi et al. [31] used the regression least-squares method to estimate the uncertain parameters of a system online and designed an indirect adaptive backstepping controller based on the estimated output value. The experimental results obtained by using this controller and a non-adaptive backstepping controller were compared. The experimental results showed that the indirect adaptive controller can monitor the expected reference signal with a slight transient behavior after parameter changes. However, non-adaptive controllers cannot keep the system on track, thereby resulting in significant oscillations and instability.

Do et al. [32] proposed an anti-winding design method to improve the performance of a strictly feedback single-input adaptive control system with input saturation to solve excessive parameter adaptation under the condition of system input saturation. The over-adaptation problem under a saturated system input was solved by appropriately modifying the adaptive law of unsaturated system inputs. Finally, the simulation results showed that this method improves the control performance in the presence of input saturation constraints.

Guan et al. [33] proposed a nonlinear adaptive robust controller with an adaptive law for a single-rod electro-hydraulic actuator with unknown nonlinear parameters. They used a specific Lyapunov function and combined the inversion method to design the controller of the entire system and the update law of all unknown parameters. Finally, a comparison of position tracking control experiments proved that the proposed nonlinear adaptive robust control method performed better than the control method without adaptive law.

Ba et al. [12] proposed a pump-controlled electro-hydraulic system integrated model inversion controller based on an integrated model inversion method to address the uncertainty and nonlinearity of a hydraulic actuator system. They used an advanced backstepping technique to compensate for nonlinearities and unknowns. Under various operating conditions, a test bench based on a pump-controlled electro-hydraulic system was used to compare the position tracking of a direct backstepping controller (DBS) and a set proportional integral derivative (PID) controller. The experimental results showed that the designed DBS exhibits high controllability.

Yao et al. [34] proposed an ideal compensation adaptive controller for noise interference caused by actual state feedback in an electro-hydraulic servo system. By considering the uncertainty of the modeling of a mismatched system and the nonlinear load pressure state, they adopted a continuous function to approximate the discontinuous symbolic function and designed a strategy on the basis of the backstepping method to construct the expected value of the intermediate state variable. Finally, their comparative experimental results validated the effectiveness of the proposed control strategy.

To address the parameter uncertainty of a nonlinear system in the position control of a pump control system (such as the oil modulus of elasticity, system leakage efficiency, and load equivalent spring stiffness), we propose in this study an adaptive backstepping control strategy for a pump control hydraulic system based on the backstepping method, and its control performance was verified experimentally.

2. Operating Principle of Pump Control System

The structure of the pump control system investigated in this study is shown in Figure 1. The system adopts the integrated volume control scheme of the servo motor and comprises a fixed displacement pump and a hydraulic cylinder. The system has a small equipment volume, simple pipeline layout, no throttle overflow loss, high reliability, high safety, and high precision. In this system, a servo motor is used to drive a fixed

displacement pump axially. The inlet port and oil drain of the fixed displacement pump are directly connected to the two load ports of the hydraulic cylinder. The accumulator is combined with a check valve to fill the oil for the system. The relief valve is used as a safety valve to protect the system pressure from exceeding the safety limit value [25]. The controller outputs torque and speed instructions to the servo motor to adjust the output pressure and flow of the fixed displacement pump, thereby enabling the output force and displacement of the hydraulic cylinder piston to be controlled.



Figure 1. Schematic diagram of electro-hydraulic servo closed-pump control system. 1: servo motor; 2: hydraulic pump; 3: oil plug; 4.1/4.2: safety valve; 5: accumulator; 6.1/6.2. the relief valve; 7.1/7.2: solenoid directional valve; 8: hydraulic cylinder. A and B represent the two chambers of the oil cylinder respectively.

The control mode of a pump control system can be typically classified into two categories: variable displacement and variable speed control. The variable speed control mode was adopted in this study to realize the position control of the pump control system by controlling the input voltage of the servo motor to adjust the output speed in real time.

3. Mathematical Model

The mathematical models of the key hydraulic components based on the pump control system are described.

3.1. Servo Motor

The AC servo motor transforms the control input voltage into the motor output speed. As the servo motor has a high response speed and a fast dynamic speed, the relationship between the motor output speed and input control signal is assumed to be a proportional loop, which is expressed as follows:

$$\omega_p = K_m u_c \tag{1}$$

where ω_p is the output speed of the motor, K_m is the control gain, and u_c is the input voltage signal.

3.2. Fixed Displacement Pump

The flow distribution characteristics of the fixed displacement pump were analyzed considering the oil compression as well as the internal and external leakage. The two-cavity load volume flow from the pump to the controlled hydraulic cylinder can be expressed as follows:

$$\begin{cases} q_B = D_p \omega_p - C_{ip} (p_A - p_B) + C_{ep} p_B \\ q_A = D_p \omega_p - C_{ip} (p_A - p_B) - C_{ep} p_A \end{cases}$$
(2)

where q_B is the flow in cavity B of the system, q_A the flow in cavity A of the system, p_B the pressure in cavity B of the system, p_A the pressure in cavity A of the system, D_p the displacement of the fixed displacement pump, ω_p the input speed of the fixed displacement pump, C_{ip} the internal leakage coefficient of the fixed displacement pump, and C_{ep} the external leakage coefficient of the fixed displacement pump.

3.3. Double-Acting Symmetrical Hydraulic Cylinder

Considering the load condition, oil compression, internal and external leakage, and other factors, we analyzed the flow distribution characteristics of the hydraulic cylinder, and the flow rate continuation equation of the two cavities of the hydraulic cylinder was established as follows:

$$\begin{cases} q_A = A_c \dot{x}_c + C_{ic} (p_A - p_B) + C_{ec} p_A + \frac{V_{cA}}{\beta_e} \dot{p}_A \\ q_B = A_c \dot{x}_c + C_{ic} (p_A - p_B) - C_{ec} p_B - \frac{V_{cB}}{\beta_e} \dot{p}_B \end{cases}$$
(3)

where A_c is the efficient working area of the hydraulic cylinder, x_c the displacement of the hydraulic cylinder, β_e the effective bulk modulus, V_{cA} the compression volume of the hydraulic cylinder's cavity A, V_{cB} the compression volume of the hydraulic cylinder's cavity B, C_{ic} the internal leakage coefficient of the hydraulic cylinder, and C_{ec} the external leakage coefficient of the hydraulic cylinder.

The force balance equation of the hydraulic cylinder is as follows:

$$A_c(p_A - p_B) = m_c \ddot{x}_c + B_c \dot{x}_c + K x_c + \tilde{d}$$
⁽⁴⁾

where m_c is the total mass converted from the load to the piston, B_c the viscous damping coefficient of oil, K the equivalent spring stiffness of the load, and \tilde{d} the external interference and unmodeled friction force.

The block diagram of the closed-pump control system can be established using the Laplace transformation expressed in Equations (1)–(4), as shown in Figure 2. On the basis of the analysis presented in [35], Equation (3) can be simplified as follows:

$$\dot{p}_{A} - \dot{p}_{B} = \frac{4\beta_{e}}{V_{t}} \left[D_{p}\omega_{p} - A_{c}\dot{x}_{c} - C_{t}(p_{A} - p_{B}) \right]$$
(5)

where V_t is the total compression volume.

Substituting the Laplace transformation expressed in Equations (1), (2) and (4) into Equation (5) yields.

$$X_{c} = \frac{A_{c}D_{p}K_{m}u_{c} - \left(C_{t} + \frac{V_{t}}{4\beta_{e}}s\right)\tilde{d}}{\frac{V_{t}}{4\beta_{e}}m_{c}s^{3} + \left(\frac{V_{t}}{4\beta_{e}}B_{c} + C_{t}m_{c}\right)s^{2} + \left(\frac{V_{t}}{4\beta_{e}}K + A_{c}^{2} + C_{t}B_{c}\right)s + C_{t}K}$$
(6)

where $C_t = C_{tc} + C_{tp} = C_{ic} + \frac{1}{2}C_{ec} + C_{ip} + \frac{1}{2}C_{ep}$.

Next, the system state variable *x* and system output *y* are defined as follows:

$$\begin{cases} \vec{x} = [x_1, x_2, x_3]^T = [x_c, \dot{x}_c, (p_A - p_B)]^T \\ y = x_1 \end{cases}$$
(7)



Figure 2. Block diagram of pump control system.

On the basis of Equations (1)–(4), the state equation of the pump control system is written as follows:

$$\begin{cases} x_1 = x_2 \\ \dot{x}_2 = -\frac{K}{m_c} x_1 - \frac{B_c}{m_c} x_2 + \frac{A_c}{m_c} x_3 + \tilde{d} \\ \dot{x}_3 = -4 \frac{\beta_e A_c}{V_t} x_2 - 4 \frac{\beta_e C_t}{V_t} x_3 + 4 \frac{\beta_e D_p K_m}{V_t} u_c \end{cases}$$
(8)

Subsequently, the following parameters are defined: $\theta_1 = K/m_c$, $\theta_2 = B_c/m_c$, $\theta_3 = A_c/m_c$, $\theta_4 = 4\beta_e A_c/V_t$, $\theta_5 = 4\beta_e C_t/V_t$, and $\theta_6 = 4\beta_e D_p K_m/V_t$. Therefore, Equation (8) can be expressed as follows:

$$\begin{cases} x_1 = x_2 \\ \dot{x}_2 = -\theta_1 x_1 - \theta_2 x_2 + \theta_3 x_3 + \tilde{d} \\ \dot{x}_3 = -\theta_4 x_2 - \theta_5 x_3 + \theta_6 u_c \end{cases}$$
(9)

4. Controller Design

On the basis of the effects of the nonlinear and uncertain parameters of the pumpcontrolled system on the position control, we used the system dynamics model presented in the previous section to design the bounded controller input of the servo motor. Figure 3 shows the control framework of the pump-control system. On the basis of the backstepping criterion, the controller selects the appropriate Lyapunov function and generates the intermediate virtual control variables step by step to obtain the adaptive backstepping controller. The backstepping method ensures that the designed adaptive controller is always Lyapunov asymptotic stable [36].



Figure 3. Control block diagram of pump control system.

The controller is designed to achieve $\lim_{t\to\infty} (x_m - x_c) = \lim_{t\to\infty} e_1 = 0$. We assume that the expected trajectory, output speed, acceleration, and acceleration derivative exist and are bounded. The design process of the controller based on backstepping is as follows:

Step 1: Define the position tracking error e_1 as

$$e_1 = x_1 - x_{1d} \tag{10}$$

where x_{1d} is the expected value of state variable x_1 . Subsequently, the derivative of e_1 with respect to time is as follows:

$$\dot{e}_1 = \dot{x}_1 - \dot{x}_{1d} = x_2 - \dot{x}_{1d} \tag{11}$$

Next, e_2 is defined as the deviation between state variable x_2 and its virtual control variable x_{2d} .

$$e_2 = x_2 - x_{2d} \tag{12}$$

An alternate Lyapunov function is defined as follows:

$$V_1 = \frac{1}{2}e_1^2 \tag{13}$$

Subsequently, the derivative of V_1 with respect to time is as follows:

$$\dot{V}_1 = e_1 \dot{e}_1 = e_1 \left(e_2 + x_{2d} - \dot{x}_{1d} \right)$$
 (14)

On the basis of Equation (11), when $x_{2d} = \dot{x}_{1d} - k_1e_1 - e_2$, the time derivative of V_1 is $\dot{V}_1 = -k_1e_1^2 \le 0$. However, because e_2 is the deviation between x_2 and x_{2d} , x_{2d} cannot contain e_2 . Hence, x_{2d} is defined as

$$x_{2d} \triangleq \dot{x}_{1d} - k_1 e_1 \tag{15}$$

where k_1 is a positive constant. Substituting Equation (15) into Equation (14) yields:

$$\dot{V}_1 = e_1 e_2 - k_1 {e_1}^2 \tag{16}$$

The aforementioned formula indicates that when e_2 is 0, V_1 is negative semidefinite, and the tracking error e_1 converges to zero. Therefore, the next task is to design a control law such that the value of e_2 approaches zero or is minimized.

Step 2: Take the derivative of x_{2d} with respect to time.

$$\dot{x}_{2d} = \ddot{x}_{1d} - k_1 \left(x_2 - \dot{x}_{1d} \right) = \ddot{x}_{1d} + k_1 \dot{x}_{1d} - k_1 x_2 \tag{17}$$

Subsequently, take the derivative of e_2 with respect to time.

$$\dot{e}_2 = \dot{x}_2 - \dot{x}_{2d} = -\theta_1 x_1 - \theta_2 x_2 + \theta_3 x_3 + \widetilde{d} - \ddot{x}_{1d} - k_1 \dot{x}_{1d} + k_1 x_2$$
(18)

Define the parameter error $\hat{\theta}_i = \hat{\theta}_i - \theta_i$, where $\hat{\theta}_i$ is an estimate of parameter θ_i .

$$\dot{e}_2 = -\left(\hat{\theta}_1 - \widetilde{\theta}_1\right)x_1 - \left(\hat{\theta}_2 - \widetilde{\theta}_2\right)x_2 + \theta_3 x_3 + \widetilde{d} - \ddot{x}_{1d} - k_1 \dot{x}_{1d} + k_1 x_2 \tag{19}$$

Define a new alternate Lyapunov function as follows:

$$V_{2} = V_{1} + \frac{1}{2}e_{2}^{2} + \frac{1}{2\alpha_{1}}\widetilde{\theta}_{1}^{2} + \frac{1}{2\alpha_{2}}\widetilde{\theta}_{2}^{2}$$
(20)

where α_i is the adaptive gain of parameter θ_i . Hence, the derivative of V_2 with respect to time is expressed as follows:

$$\dot{V}_{2} = \dot{V}_{1} + e_{2}\dot{e}_{2} + \frac{1}{\alpha_{1}}\widetilde{\theta}_{1}\hat{\theta}_{1} + \frac{1}{\alpha_{2}}\widetilde{\theta}_{2}\hat{\theta}_{2}
= -k_{1}e_{1}^{2} + e_{2}\left[e_{1} - \hat{\theta}_{1}x_{1} - \hat{\theta}_{2}x_{2} + \theta_{3}x_{3} - \ddot{x}_{1d} - k_{1}\dot{x}_{1d} + k_{1}x_{2}\right]
+ e_{2}\widetilde{d} + e_{2}\widetilde{\theta}_{1}x_{1} + e_{2}\widetilde{\theta}_{2}x_{2} + \frac{1}{\alpha_{1}}\widetilde{\theta}_{1}\dot{\hat{\theta}}_{1} + \frac{1}{\alpha_{2}}\widetilde{\theta}_{2}\dot{\hat{\theta}}_{2}$$
(21)

Define e_3 as the deviation between state variable x_3 and its virtual control variable x_{3d} .

$$x_3 = x_3 - x_{3d} \tag{22}$$

Similarly, use Equation (21) to design virtual control variable x_{3d} .

е

$$x_{3d} = \frac{1}{\theta_3} \left[-k_2 e_2 + \ddot{x}_{1d} + k_1 \dot{x}_{1d} - k_1 x_2 + \hat{\theta} x_1 + \hat{\theta}_2 x_2 - e_1 \right]$$
(23)

Here, if $k_2 > 0$, then Equation (21) can be simplified as Equation (23).

$$\dot{V}_{2} = -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} + \theta_{3}e_{2}e_{3} + e_{2}\tilde{d} + \tilde{\theta}_{1}\left(e_{2}x_{1} + \frac{1}{\alpha_{1}}\dot{\theta}_{1}\right) + \tilde{\theta}_{2}\left(e_{2}x_{2} + \frac{1}{\alpha_{2}}\dot{\theta}_{2}\right)$$
(24)

On the basis of the aforementioned formula, when e_3 and \tilde{d} are equal to 0, \dot{V}_2 is negative semidefinite. Therefore, the next step is to ensure that e_3 and \tilde{d} are equal to zero or as small as possible.

Step 3: Because the final control variable x_3 is visualized and the control variable u_c exists on its own, virtual control variables are not necessitated in this step.

The derivative of x_{3d} with respect to time can be expressed as

. 1

$$\dot{x}_{3d} = A\theta_1 + B\theta_2 + C + Dd \tag{25}$$

where

$$\begin{cases} A &= \frac{1}{\theta_3} \left[k_2 x_1 + k_1 x_1 - \hat{\theta}_2 x_1 \right] \\ B &= \frac{1}{\theta_3} \left[k_2 x_2 + k_1 x_2 - \hat{\theta}_2 x_2 \right] \\ C &= \frac{1}{\theta_3} \left[-k_2 \left(k_1 x_2 - \ddot{x}_{1d} - k_1 \dot{x}_{1d} \right) + \dot{\hat{\theta}}_1 x_1 + \hat{\theta}_1 x_2 + \dot{\hat{\theta}}_2 x_2 \right] \\ &- x_2 + \dot{x}_{1d} + k_1 \ddot{x}_{1d} + \ddot{x}_{1d} \right] + \left(\hat{\theta}_2 - k_1 \right) x_3 - k_2 x_3 \\ D &= \frac{1}{\theta_3} \left[\hat{\theta}_2 - k_2 - k_1 \right] \end{cases}$$

Therefore,

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$$\dot{e}_3 = \theta_6 u_c - C - D\tilde{d} - A\left(\hat{\theta}_1 - \tilde{\theta}_1\right) - B\left(\hat{\theta}_2 - \tilde{\theta}_2\right) - \left(\hat{\theta}_4 - \tilde{\theta}_4\right) x_2 - \left(\hat{\theta}_5 - \tilde{\theta}_5\right) x_3$$
(26)

Next, the Lyapunov function is defined as

$$V_3 = V_2 + \frac{1}{2}e_3^2 + \frac{1}{2\alpha_4}\tilde{\theta}_4^2 + \frac{1}{2\alpha_5}\tilde{\theta}_5^2$$
(27)

Therefore, the derivative of V_3 with respect to time is expressed as follows:

$$\dot{V}_{3} = -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} + e_{3}\left[\theta_{3}e_{2} - \hat{\theta}_{4}x_{2} - \hat{\theta}_{5}x_{3} + \theta_{6}u_{c} - \hat{\theta}_{1}A - \hat{\theta}_{2}B - C - D\tilde{d}\right] + e_{2}\tilde{d} + \tilde{\theta}_{1}\left(e_{3}A + e_{2}x_{1} + \frac{1}{\alpha_{1}}\dot{\theta}_{1}\right) + \tilde{\theta}_{2}\left(e_{3}B + e_{2}x_{2} + \frac{1}{\alpha_{2}}\dot{\theta}_{2}\right) + \tilde{\theta}_{4}\left(e_{3}x_{2} + \frac{1}{\alpha_{4}}\dot{\theta}_{4}\right) + \tilde{\theta}_{5}\left(e_{3}x_{3} + \frac{1}{\alpha_{5}}\dot{\theta}_{5}\right)$$
(28)

The actual control variable u_c can be designed as follows:

$$u_{c} = \frac{1}{\theta_{6}} \left[-k_{3}e_{3} - \theta_{3}e_{2} + \hat{\theta}_{4}x_{2} + \hat{\theta}_{5}x_{3} + \hat{\theta}_{1}A + \hat{\theta}_{2}B + C + d_{n2}\operatorname{sgn}(e_{3}) - \frac{e_{2}}{e_{3}}d_{n1}\operatorname{sgn}(e_{2}) \right]$$
(29)

where $k_3 > 0$, $d_{n1} \ge \left| \tilde{d} \right|$, and $d_{n2} \ge \left| D\tilde{d} \right|$. Furthermore, the sgn(*e*) function is defined

as follows:

$$\operatorname{sgn}(e) = \begin{cases} 1, \ e > 0 \\ 0, \ e = 0 \\ -1, \ e < 0 \end{cases}$$
(30)

On the basis of Equations (28)–(30), the derivative of V_3 with respect to time can be expressed as

$$\dot{V}_{3} = -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} - k_{3}e_{3}^{2} - e_{3}\left[D\tilde{d} - d_{n2}\mathrm{sgn}(e_{3})\right] + e_{2}\left[\tilde{d} - d_{n1}\mathrm{sgn}(e_{2})\right]
+ \tilde{\theta}_{1}\left(e_{3}A + e_{2}x_{1} + \frac{1}{\alpha_{1}}\dot{\theta}_{1}\right) + \tilde{\theta}_{2}\left(e_{3}B + e_{2}x_{2} + \frac{1}{\alpha_{2}}\dot{\theta}_{2}\right)
+ \tilde{\theta}_{4}\left(e_{3}x_{2} + \frac{1}{\alpha_{4}}\dot{\theta}_{4}\right) + \tilde{\theta}_{5}\left(e_{3}x_{3} + \frac{1}{\alpha_{5}}\dot{\theta}_{5}\right)$$
(31)

The adaptive law of system parameters is expressed as follows:

$$\begin{cases} \hat{\theta}_{1} = -\alpha_{1}(e_{3}A + e_{2}x_{1}) \\ \hat{\theta}_{2} = -\alpha_{2}(e_{3}B + e_{2}x_{2}) \\ \hat{\theta}_{4} = -\alpha_{4}e_{3}x_{2} \\ \hat{\theta}_{5} = -\alpha_{5}e_{3}x_{3} \end{cases}$$
(32)

Therefore, V_3 can be simplified as

$$\dot{V}_3 = -k_1 e_1^2 - k_2 e_2^2 - k_3 e_3^2 - e_3 \left[D\tilde{d} - d_{n2} \operatorname{sgn}(e_3) \right] + e_2 \left[\tilde{d} - d_{n1} \operatorname{sgn}(e_2) \right] \le 0$$
(33)

As shown in Equation (27), the semi-positive Lyapunov function V_3 is constructed using Equations (13) and (20), separately. When we adopt Equations (29) and (32) as the input and parameter adaptive law of the system, its derivative V_3 is semi-negative definite. According to Lyapunov stability theory, control errors e_2 and e_3 are stable at the equilibrium point. According to Barbalat's lemma, when the state error e_2 converges asymptotically to zero, the system output error e_1 converges asymptotically to 0 as well.

Therefore, when the pump control system adopts the voltage control rate and the parameter adaptive law, as shown in Equations (29) and (32), the final output position error of the electro-hydraulic servo closed-pump control system will gradually converge to 0.

5. Experimental Results and Analysis

5.1. Experimental Platform

The proposed adaptive backstepping control algorithm was verified experimentally by using the electro-hydraulic servo closed-pump control system experimental platform presented in [37]. Figure 4 shows that the experimental platform is primarily composed of a power unit, a valve block, a hydraulic cylinder, and an electric control cabinet. The working and loading cylinder adopts the structure of the built-in SSI displacement sensor (the sensor model is KH10MB0060MC81S1B10, and the resolution is 1 μ m). The power unit adopts an electro-hydraulic servo pump control unit, which is composed of a radial piston pump driven by a permanent magnet synchronous servo motor manufactured by MOOG, referred to as an electro-hydraulic servo pump control unit (herein referred to as EPU). The specific structures are shown in Figure 5.



Figure 4. Test platform.



Figure 5. Electro hydraulic servo pump control unit.

Figure 6 shows the control structure of the experimental platform. The control software adopts the MOOG axis control software, MACS 3.4. The control hardware is composed of an Advantech IPC-610 computer; Beckhoff modules EK1100, EL3122, and EL2004; a host computer MOOG controller MSD; an MOOG single-axis servo driver; and other auxiliary components. The PC of the device uses MACS 3.4 to write the master control program, imports the MSD of the upper computer through TCP/IP, sends control signals to the servo driver via EtherCat communication, and receives feedback signals from the Beckhoff acquisition module simultaneously.

5.2. Experimental Analysis

The main parameters of the experimental platform are listed in Table 1. The adaptive backstepping control algorithm (ABC) was developed using the MATLAB/Simulink software and downloaded to the MOOG MACS axis control software. In this software, the system logic control program was edited to control the position output of the electrohydraulic servo pump control system. Using the opposite vertex, the cylinder simulated the elastic load of the system, and the expected target signals of the experimental platform were the "S" slope trajectory and sine wave signals, separately. The traditional PID controller and adaptive backstepping controller were used separately to perform the experiment.



Figure 6. Pump control test bench control frame.

Parameter	Symbol	Units	Value
Total compression volume	V_t	m ³	$4.15 imes 10^{-2}$
Efficient working area cylinder	A_c	m ²	$7.9 imes10^{-4}$
Total mass converted from the load to the piston	m_c	kg	150
Viscous damping coefficient	B _c	$\mathrm{N(ms^{-1})}^{-1}$	0.0345
Equivalent spring stiffness of the load	Κ	Mm^{-1}	$5 imes 10^7$
Total leakage coefficient of hydraulic system	C_t	$m^3(s \cdot Pa)^{-1}$	$9 imes 10^{-11}$
Effective volume modulus of oil	β_{e}	Nm^{-2}	$7 imes 10^8$
Control gain	K_m	(r/min)/V	300
Displacement of fixed displacement pump	D_p	m ³ /rad	$8 imes 10^{-6}$
External disturbance and unmodeled friction	d	Ν	$1.5 imes 10^4$

Table 1. Operating parameters of experimental platform.

5.2.1. 10 mm "S" Ramp Signal

Taking the working load of the pump oil control engine of the thermal power turbine set as an example, we set the load to simulate the output of the hydraulic cylinder of 5 kN and the load fluctuation is ± 0.5 kN. By providing an "S"-type ramp signal of 10 mm displacement under the load condition of the system, we obtained the experimental curves of position extension and retraction, as shown in Figures 7 and 8.



Figure 7. Cont.



Figure 7. Cont.



Figure 7. The 40–50 mm displacement tracking curve. (a) Position trailing; (b) Tracking error; (c) Motor speed; (d) Pressure difference.



Figure 8. Cont.



Figure 8. Cont.



Figure 8. The 50–40 mm displacement tracking curve. (a) Position trailing; (b) Tracking error; (c) Motor speed; (d) Pressure difference.

Figure 7 shows that the following error of using ABC was much smaller than that of using traditional PID. The steady-state control accuracy of ABC was able to reach ± 0.02 mm, while the PID steady-state control accuracy was only ± 0.05 mm. Moreover, the time for the system using ABC to reach the steady state was shorter.

As shown in Figure 8, the ABC designed in this paper had better tracking performance than the traditional PID, which is beneficial in terms of alleviating chattering and overshooting in the control process. In addition, the overshoot and the time to reach the steady state of the system using ABC were far less than the time to reach the steady state of the system using PID.

5.2.2. Sine Signal

To further compare and analyze the dynamic characteristics of the traditional PID and ABC, we imposed sine signals with an amplitude of 0.5 mm and a frequency of 0.6 Hz at 50 mm; the experimental results are shown in Figure 9.

As shown in Figure 9, ABC presents better dynamic response characteristics than the traditional PID.

The tracking error of the ABC control was less than ± 0.04 mm, and the tracking error of the traditional PID was approximately ± 0.06 mm.

The aforementioned experiments indicated that the adaptive backstepping control strategy proposed herein exhibits good control performance in practical engineering applications.



Figure 9. Sinusoidal tracking curve. (a) Position trailing; (b) Tracking error.

6. Conclusions

Aiming at the position control problem of electro-hydraulic servo closed pump control system, we proposed an adaptive backstepping control strategy. Conclusions are as follows:

- (1) The mathematical model of the electro-hydraulic servo pump control system was established, and the position output transfer function of the system was deduced.
- (2) An adaptive backstepping control strategy was proposed on the basis of the backstepping method. The algorithm fully considers the nonlinearity and parameter uncertainty of the pump control system. When the desired control input is obtained, the adaptive adjustment rate of the uncertain parameter is derived and applied to actual position control.
- (3) Experimental analysis showed that the adaptive backstepping control strategy proposed in this paper had good control performance in practical applications. Its steady-state control accuracy was able to reach ± 0.02 mm, which can lay a certain foundation for high-precision position control of the pump control system. In addition, the system dead zone characteristics are also the main factors affecting the

steady-state control. In order for the control accuracy of the system to be further improved, the pump control system dead zone characteristics can be further studied.

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Nomenclature

- DBS direct backstepping controller
- PID proportional integral derivative controller
- EPU electro-hydraulic servo pump control unit
- ABC adaptive backstepping control algorithm

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